

Fuji IGBT Module

Application Manual

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Chapter 2 Terms and Characteristics

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2-1



This chapter describes the terms and characteristics of IGBT modules.

1. Explanation of Terms

Various terms used in the specifications are explained below.

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Term	Symbol	Definition explanation (Refer to specifications for test conditions)
Collector-Emitter voltage	V _{CES}	Maximum Collector-Emitter (hereinafter referred to as C-E) voltage with Gate-Emitter (G-E) shorted.
Gate-Emitter voltage	V _{GES}	Maximum G-E voltage with C-E shorted.
Collector current	<i>I</i> _C	Maximum DC collector current of IGBT.
	I _{CRM} I _C pulse	Maximum pulse collector current of IGBT.
Forward current	- <i>I</i> _C	Maximum DC forward current of FWD.
	I _{FRM} -I _C pulse	Maximum pulse forward current of FWD.
Total power dissipation	P _{tot} P _C	Maximum power dissipation per IGBT.
Virtual junction temperature	T _{vj} Tj	Maximum junction temperature at which normal operation is possible. You must not exceed this temperature under the worst condition.
Operating virtual junction temperature	T _{vj(op)} T _{vjop}	Junction temperature during continuous operation.
Case temperature	T _C	Case temperature directly below the IGBT chip or FWD chip that generates the most heat inside the IGBT module. (refer to Chapter 6 for details.)
Storage temperature	T _{stg}	Temperature range allowing storage or transportation without being subjected to electrical load.
FWD ^P t	ft	Joule-integral value of overcurrent allowed that does not result in diode destruction. The overcurrent is defined by a line frequency half sine wave (50, 60Hz) at one cycle.
FWD surge forward current	I _{FSM}	The maximum value of overcurrent allowed that does not result in diode destruction. The overcurrent is defined by a line frequency half sine wave (50, 60Hz).
Isolation voltage	V _{iso}	Maximum sinusoidal voltage RMS value allowed between all shorted terminals and the heat sink mounting surface.
	Mounting	Maximum allowable torque value when mounting the IGBT module on a heat sink with the specified screws
Screw torque		Maximum allowable torque value when connecting external busher or

Table 2-1 Maximum ratings

Note 1: Values at $T_{\rm C}$ = 25 °C unless otherwise specified

Terminal

Note 2: The values listed as maximum ratings should not be exceeded under any circumstances.

wires to the terminals with the specified screws

Maximum allowable torque value when connecting external busbar or



	Term	Symbol	Definition explanation (Refer to specifications for test conditions)
Static characteristics	Zero gate voltage collector current	ICES	Collector current when a specific voltage is applied between C-E with G-E shorted.
	Gate-Emitter leakage current	I _{GES}	Gate current when a specific voltage is applied between G-E with C-E shorted.
	Gate-Emitter threshold voltage	V _{GE(th)}	G-E voltage (V_{GE}) at a specified I_C and C-E voltage.
	Collector-Emitter saturation voltage	V _{CE(sat)}	C-E voltage at a specified collector current and G-E voltage.
	Input capacitance	Cies	G-E capacitance when a specified voltage is applied between G-E and C-E while C-E is shorted in AC.
	Output capacitance	Coes	C-E capacitance when a specified voltage is applied between G-E and C-E while G-E is shorted in AC.
	Reverse transfer capacitance	Cres	C-G capacitance when a specified voltage is applied between G-E and C-E while G-E and C-E are shorted in AC.
	Forward voltage	V _F	Forward voltage of FWD at a specified forward current.
Dynamic characteristics	Turn-on time	ton	The time during IGBT turn-on when V_{GE} rises from 10% of the maximum value until collector current rises to 90% of the maximum value.
	Rise time	<i>t</i> r	The time during IGBT turn-on when the collector current rises from 10% to 90% of the maximum value.
	Turn-off time	t _{off}	The time during IGBT turn-off when V_{GE} drops from 10% of the maximum value until collector current drops to 10% of the maximum value.
	Fall time	t _f	The time during IGBT turn-off when the collector current drops from 90% to 10% of the maximum value.
	Turn-on loss	Eon	Loss generated during IGBT turn-on.
	Turn-off loss	E _{off}	Loss generated during IGBT turn-off.
	Reverse recovery loss	E _{rr}	Loss generated during FWD reverse recovery.
	Reverse recovery time	<i>t</i> _{rr}	Time required for FWD reverse recovery current to disappear.
	Reverse recovery current	I _{rr} (I _{rp})	Peak reverse current during reverse recovery.
Reverse bias safe operating area RBSOA		RBSOA	Current and voltage range where IGBT can be turned off safely under specified conditions.
Internal gate resistance $\frac{r_{g}}{R_{g(int)}}$		r_{g} $R_{g(int)}$	Built-in gate series resistance.
Gate	e charge	Qg	Amount of charge required to turn-on the IGBT.

Table 2-2 Electrical characteristics

Note 1: The definition of dynamic characteristics differs between series. For details, refer to the specifications of each product.



Table 2-3 Thermal resistance characteristics

Term	Symbol	Definition explanation (Refer to specifications for test conditions)
Thermal resistance	R _{th(j-c)}	Thermal resistance between the chip (junction)and case of IGBT or FWD.
	$R_{ m th(c-s)}$	Thermal resistance between the case and heat sink when the IGBT module is mounted on a heat sink with the recommended torque and thermal grease.

Table 2-4 Thermistor characteristics

Term	Symbol	Definition explanation (Refer to specifications for test conditions)
Thermistor resistance	R	Thermistor resistance at specified temperature
B value	В	Value indicating the magnitude of thermistor resistance change between any two temperatures in resistance-temperature characteristics.



2. Characteristics of IGBT and FWD

Using the 6MBI100VB-120-50 (1200V/100A, 6th generation IGBT module) as an example, the various characteristics of the IGBT described in the specifications are explained below.

2.1 Static characteristics

Fig. 2-1 and Fig. 2-2 show the V_{GE} dependence of V_{CE} - I_C characteristics (output characteristics). While the IGBT is on, V_{CE} changes in accordance with I_C , V_{GE} , and T_{vj} . V_{CE} represents the C-E voltage drop during the ON state, and is used to calculate the power dissipation loss of the IGBT. The smaller the V_{CE} value, the lower the power dissipation loss.

It is generally recommended to operate at V_{GE} =15V, and the collector current lower than the module current rating.

Fig. 2-3 shows the data in Fig. 2-1 as $I_{\rm C}$ dependence of $V_{\rm CE}$ - $V_{\rm GE}$ characteristics. From the graph, the $V_{\rm GE}$ value limit where $V_{\rm CE}$ suddenly increases can be read.







Fig. 2-2 $V_{CE(sat)}$ - I_C characteristics (T_{vj} =150°C)



Fig. 2-3 V_{CE} - V_{GE} characteristics (T_{vi} =25°C)



2.2 Switching characteristics

As the IGBT is generally used for switching applications, it is important to fully understand the turnon and turn-off switching characteristics. Since these characteristics are affected by various parameters, it is necessary to take these into consideration when determining operating conditions.

Switching characteristics can be divided into switching time and switching loss. These switching characteristics can be measured by the chopper circuit shown in Fig. 2-4.

The definitions of switching times are shown in Fig. 2-5.



Fig. 2-4 Switching characteristics measuring circuit



Fig. 2-5 Definition of switching time

2-6



The relationship between switching time and collector current $I_{\rm C}$ is shown in Fig. 2-6 and Fig. 2-7. Fig. 2-8 shows the relationship between switching time and gate resistance $R_{\rm G}$. As shown in these figures, the switching time varies depending on $I_{\rm C}$, $T_{\rm vj}$, and $R_{\rm G}$, so please take this into consideration when designing equipment.

For example, when the IGBT is used under the condition that t_{off} is too long, it may exceed the dead time and cause a series arm short-circuit.

It is also important to be aware that if the switching time, such as t_f is too short, the transient current change rate dI_c/dt will increase and may cause a high turn-off surge voltage (= $L_S \cdot dI_C/dt$). Since this surge voltage is superimposed on the applied voltage, the resulting voltage might exceed RBSOA and destroy the IGBT (refer to Chapter 2 and 4 for details).



Fig. 2-6 Switching time - $I_{\rm C}$ characteristics ($T_{\rm vi}$ =125°C)



Fig. 2-7 Switching time - $I_{\rm C}$ characteristics ($T_{\rm vi}$ =150°C)



Fig. 2-8 Switching time - $R_{\rm G}$ characteristics $(T_{\rm vi}$ =125°C)



Switching loss (E_{on} , E_{off} , E_{rr}) occurs every time an IGBT is turned on or off, therefore it is important to minimize this loss as much as possible. As shown in Fig. 2-9 and 2-10, switching loss changes in accordance with I_{C} , T_{vj} , and R_{G} . In particular, the selection of R_{G} is important. If it is too large, the switching loss will increase, and series arm short circuit due to the aforementioned insufficient dead time will easily occur. Conversely, if R_{G} is reduced to reduce switching loss, the aforementioned excessive surge voltage (= L_{S} ·d I_{C} /dt) may occur.

As can be seen from this, the value of main circuit inductance L_S has great influence on R_G selection. The smaller the value is, the smaller the surge voltage will be, making it easier to consider R_G selection. Therefore, it is recommended to design the L_S value as small as possible.

When selecting R_G , it is also necessary to consider matching with the capacitance of the IGBT drive circuit. Therefore, please select the R_G after careful consideration using the capacitance characteristics as shown in Chapter 2.3.



Fig. 2-9 Switching loss - I_C characteristics



Fig. 2-10 Switching loss - R_G characteristics



2.3 Capacitance characteristics

Fig. 2-11 shows the characteristics of Q_g . This characteristic shows the change of V_{CE} , V_{GE} with respect to Q_g . ' Q_g increases' means 'charging the G-E capacitance of the IGBT', so when Q_g increases, V_{GE} (= Q_g / G-E capacitance) rises and the IGBT turns on. When the IGBT turns on, V_{CE} drops to the saturation voltage. Thus, Q_g indicates the amount of charge required to drive the IGBT. Use this characteristic when determining the power supply capacity of the drive circuit.

Fig. 2-12 shows the IGBT junction capacitance characteristics. Fig. 2-13 shows the junction capacitance C_{ies} , C_{oes} and C_{res} . Use these characteristics along with Q_{g} to design your drive circuit.



Fig. 2-11 V_{CE} , V_{GE} - Q_g characteristics



Fig. 2-12 Junction capacitance characteristic



Fig. 2-13 Junction capacitance



2.4 Reverse biased safe operating area

During turn-off, IGBT has a safe operating area defined by V_{CE} an I_C called RBSOA (Reverse Bias Safe Operating Area). Fig. 2-14 shows the area of 1200V/100A IGBT as an example.

It is important to design the snubber circuit that will keep V_{CE} and I_C within the region of RBSOA when the IGBT is turned off. In the case of a short circuit (non-repetitive), the IGBT safe operating area defined by V_{CE} an I_C is called SCSOA (Short Circuit Safe Operating Area). SCSOA is different for each IGBT series, thus refer to the technical data of each series for details.



Fig. 2-14 RBSOA

2.5 FWD characteristics

The IGBT module has built-in anti-parallel FWD. Fig. 2-15 shows the FWD $V_{\rm F}$ - $I_{\rm F}$ characteristic, and Fig. 2-16 shows the reverse recovery characteristic ($t_{\rm rr}$, $I_{\rm rr}$). $E_{\rm rr}$ characteristics are shown in Fig. 2-9 and 2-10. Use these characteristics to calculate the power loss of FWD as well as the IGBT. Note that these characteristics change in accordance with $I_{\rm F}$, $T_{\rm vi}$, and $R_{\rm G}$.



Fig. 2-15 $V_{\rm F}$ - $I_{\rm F}$ characteristics



Fig. 2-16 $t_{\rm rr}$, $I_{\rm rr}$ - $I_{\rm F}$ characteristics



2.6 Transient thermal resistance characteristics

Fig. 2-17 shows the transient thermal resistance characteristics used for temperature rise calculation and heat sink design (this characteristic is for one arm for both IGBT and FWD).

This thermal resistance is a characteristic often used in thermal analysis, and the formula is very similar to Ohm's law of electrical resistance. It is defined as:

Temperature difference ΔT [°C] = Thermal resistance R_{th} [°C / W] x Energy E (loss) [W]

In the IGBT module, the thermal resistance is used when calculating the T_{vj} of the IGBT and FWD. (For details, refer to Chapter 6 'Cooling Design')



Fig. 2-17 Transient thermal resistance